

# OPA727, OPA2727 OPA4727, OPA728

SBOS314H-SEPTEMBER 2004-REVISED APRIL 2007

# e-trim<sup>™</sup> 20MHz, High Precision CMOS Operational Amplifier

#### **FEATURES**

OFFSET: 15μV (typ), 150μV (max)
 DRIFT: 0.3μV/°C (typ), 1.5μV/°C (max)

BANDWIDTH: 20MHz
 SLEW RATE: 30V/μs

BIAS CURRENT: 500pA (max)
 LOW NOISE: 6nV/√Hz at 100kHz

• THD+N: 0.0003% at 1kHz

QUIESCENT CURRENT: 4.3mA/ch
 SUPPLY VOLTAGE: 4V to 12V
 SHUTDOWN MODE (OPA728): 6μA

#### **APPLICATIONS**

OPTICAL NETWORKING

• TRANSIMPEDANCE AMPLIFIERS

INTEGRATORS

• ACTIVE FILTERS

A/D CONVERTER DRIVERS

I/V CONVERTER FOR DACs

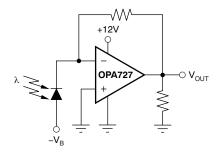
HIGH PERFORMANCE AUDIO

PROCESS CONTROL

TEST EQUIPMENT

#### **OPAX727 AND OPAX728 RELATED PRODUCTS**

FEATURES	PRODUCT
20MHz, 3mV, 4μV/°C (non-e-trim version of OPA727)	OPA725
20MHz, 3mV, 4μV/°C, Shutdown (non-e-trim version of OPA728)	OPA726



#### DESCRIPTION

The OPA727 and OPA728 series op amps use a state-of-the-art 12V analog CMOS process and e-trim, a package-level trim, offering outstanding dc precision and ac performance. The extremely low offset (150 $\mu$ V max) and drift (1.5 $\mu$ V/°C) are achieved by trimming the IC digitally after packaging to avoid the shift in parameters as a result of stresses during package assembly. To correct for offset drift, the OPA727 and OPA728 family is trimmed over temperature. The devices feature very high CMRR and open-loop gain to minimize errors.

Excellent ac characteristics, such as 20MHz GBW, 30V/µs slew rate and 0.0003% THD+N make the OPA727 and OPA728 well-suited for communication, high-end audio, and active filter applications. With a bias current of less than 500pA, they are well suited for use as transimpedance (I/V-conversion) amplifiers for monitoring optical power in ONET applications.

Optimized for single-supply operation up to 12V, the input common-mode range extends to GND for true single-supply functionality. The output swings to within 150mV of the rails, maximizing dynamic range. The low quiescent current of 4.3mA makes it well-suited for use in battery-operated equipment. The OPA728 shutdown version reduces the quiescent current to typically 6µA and features a reference pin for easy shutdown operation with standard CMOS logic in dual-supply applications.

For ease of use, the OPA727 and OPA728 op amp families are fully specified and tested over the supply range of 4V to 12V. The OPA727 (single) and OPA728 (single with shutdown) are available in MSOP-8 and DFN-8; the OPA2727 (dual) is available in DFN-8 and SO-8; and the quad version OPA4727 in TSSOP-14. All versions are specified for operation from -40°C to +125°C.

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### PACKAGE/ORDERING INFORMATION(1)

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	PACKAGE MARKING
Non-Shutdown			
OPA727	MSOP-8	DGK	AUE
OPA727	DFN-8	DRB	NSF
OD40707	DFN-8	DRB	NSD
OPA2727	SO-8	D	O2727A
OPA4727	TSSOP-14	PW	OPA4727
Shutdown			
ODA700	MSOP-8	DGK	AUF
OPA728	DFN-8	DRB	NSG

<sup>(1)</sup> For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

#### ABSOLUTE MAXIMUM RATINGS(1)

		OPA727, OPA2727 OPA4727, OPA728	UNIT
Supply Voltage		+13.2	V
Cianal Innut Tarminala	Voltage <sup>(2)</sup>	-0.5 to (V+) + 0.5	V
Signal Input Terminals	Current <sup>(2)</sup>	±10	mA
Output Short-Circuit (3)		Continuous	
Operating Temperature		-55 to +125	°C
Storage Temperature		-55 to +150	°C
Junction Temperature		+150	°C
ECD Detice	Human Body Model	2000	V
ESD Rating	Charged Device Model	1000	V

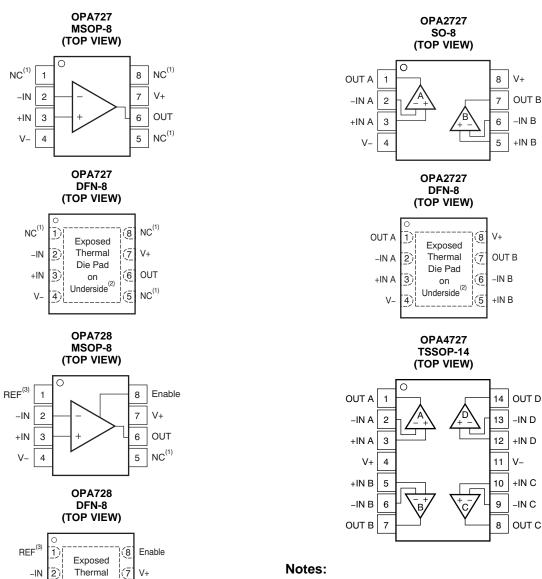
<sup>(1)</sup> Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not supported.

<sup>(2)</sup> Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current limited to 10mA or less.

<sup>(3)</sup> Short-circuit to ground, one amplifier per package.



#### **PIN CONFIGURATIONS**



Die Pad

οn

Underside(2)

+IN3) (ē

OUT

 $NC^{(1)}$ 

- 1. NC denotes no internal connection.
- 2. Connect thermal die pad to V-.
- 3. REF is the reference voltage for ENABLE pin.



### ELECTRICAL CHARACTERISTICS: $V_s$ = +4V to +12V or $V_s$ = ±2V to ±6V

**Boldface** limits apply over the specified temperature range,  $T_A = -40^{\circ}C$  to  $+125^{\circ}C$ . At  $T_A = +25^{\circ}C$ ,  $R_L = 10k\Omega$  connected to  $V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.

PARAMETER			O OF			
		CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET VOLTAGE						
Input Offset Voltage	Vos	$V_S = \pm 5V$ , $V_{CM} = 0V$				
OPA727 DFN, OPA728 DFN Packages				15	150	μV
OPA727 MSOP, OPA728 MSOP Packages				15	300	μV
OPA2727				15	150	μV
OPA4727				15	175	μV
Drift	$\mathrm{dV}_{\mathrm{OS}}/\mathrm{dT}$	0°C to +85°C		0.3	1.5	μV/°C
		-40°C to +125°C		0.6	3	μ <b>V</b> /° <b>C</b>
vs Power Supply	PSRR	$V_S = \pm 2V$ to $\pm 6V$ , $V_{CM} = V$		30	150	μV/V
Over Temperature		$V_S = \pm 2V$ to $\pm 6V$ , $V_{CM} = V-$			150	μ <b>V/V</b>
Channel Separation, dc				1		μV/V
INPUT BIAS CURRENT						
Input Bias Current				±85	±500	pA
Over Temperature			See T	ypical Charac	cteristics	
Input Ofset Current	Ios			±10	±100	pA
NOISE						
Input Voltage Noise, f = 0.1Hz to 10Hz	e <sub>n</sub>	$V_{S} = \pm 6V, V_{CM} = 0V$		10		$\mu V_{PP}$
Input Voltage Noise Density, f = 10kHz	en	$V_{S} = \pm 6V, V_{CM} = 0V$		10		nV/√ <del>Hz</del>
Input Voltage Noise Density, f = 100kHz	en	$V_{S} = \pm 6V, V_{CM} = 0V$		6		nV/√ <del>Hz</del>
Input Current Noise Density, f = 1kHz	i <sub>n</sub>	$V_{S} = \pm 6V, V_{CM} = 0V$		2.5		fA/√ <del>Hz</del>
INPUT VOLTAGE RANGE						
Common-Mode Voltage Range	$V_{CM}$		(V-)		(V+)-2.5	V
Common-Mode Rejection Ratio	CMRR	$(V-) \le V_{CM} \le (V+) - 2.5V$	86	94		dB
Over Temperature		$(V-) \le V_{CM} \le (V+) - 2.5V$	84			dB
		$(V-) \le V_{CM} \le (V+) - 3V$	94	100		dB
Over Temperature		$(V-) \le V_{CM} \le (V+) - 3V$	84			dB
INPUT IMPEDANCE						
Differential				10 <sup>11</sup>    5		Ω    pF
Common-Mode				1011    4		Ω    pF
OPEN-LOOP GAIN				"		
Open-Loop Voltage Gain	A <sub>OL</sub>	$R_L = 100k\Omega$ , $0.15V < V_O < (V+)$ -0.15V	110	120		dB
Over Temperature		$R_L = 100 k\Omega, 0.15 V < V_O < (V+)$ -0.15 V	100			dB
		$R_L = 1k\Omega$ , $0.25V < V_O < (V+) -0.25V$	106	116		dB
Over Temperature, OPA727, OPA728		$R_L = 1k\Omega$ , 0.25V < $V_O$ < (V+) -0.25V	96			dB
Over Temperature, OPA2727, OPA4727		$R_L = 1 k\Omega$ , $0.35 V < V_O < (V+) -0.35 V$	96			dB
FREQUENCY RESPONSE	-	C <sub>L</sub> = 20 pF				
Gain-Bandwidth Product	GBW			20		MHz
Slew Rate	SR	G = +1		30		V/μs
Settling Time, 0.1%	$t_s$	$V_S = \pm 6V$ , 5V Step, $G = +1$		350		ns
0.01%		$V_S = \pm 6V$ , 5V Step, $G = +1$		450		ns
Overload Recovery Time		V <sub>IN</sub> × Gain > V <sub>S</sub>		50		ns
Total Harmonic Distortion + Noise	THD+N	$V_S = \pm 6V$ , $V_{OUT} = 2V_{RMS}$ , $R_L = 600\Omega$ ,		0.003		%
		G = +1, f = 1kHz				



### ELECTRICAL CHARACTERISTICS: $V_s$ = +4V to +12V or $V_s$ = ±2V to ±6V (continued)

**Boldface** limits apply over the specified temperature range,  $T_A = -40^{\circ}C$  to  $+125^{\circ}C$ . At  $T_A = +25^{\circ}C$ ,  $R_L = 10k\Omega$  connected to  $V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.

PARAMETER				OPA727, OPA728, OPA2727, OPA4727				
		CONDITIONS	MIN	TYP	MAX	UNIT		
OUTPUT								
Voltage Output Swing from Rail								
		$R_L = 100k\Omega$ , $A_{OL} > 110dB$		100	150	mV		
Over Temperature		$R_L = 100k\Omega$ , $A_{OL} > 100dB$			150	mV		
		$R_L = 1k\Omega, A_{OL} > 106dB$		200	250	mV		
Over Temperature, OPA727, OPA728		$R_L = 1k\Omega$ , $A_{OL} > 96dB$			250	mV		
Over Temperature, OPA2727, OPA4727		$R_L = 1k\Omega$ , $A_{OL} > 96dB$			350	mV		
Output Current	I <sub>OUT</sub>	$ V_S - V_{OUT}  < 1V$		40		mA		
Short-Circuit Current	I <sub>SC</sub>			±55		mA		
Capacitive Load Drive	$C_{LOAD}$		See Ty	pical Chara	acteristics			
Open-Loop Output Impedance		$f = 1MHz, I_O = 0$		40		Ω		
ENABLE/SHUTDOWN (OPA728)								
t <sub>OFF</sub>				5		μs		
t <sub>ON</sub>				80		μs		
Enable Reference (Ref Pin) Voltage Range			V-		(V+) -2	V		
V <sub>L</sub> (amplifier is disabled)					< V <sub>DGND</sub> +0.8V	V		
V <sub>H</sub> (amplifier is enabled)			> V <sub>DGND</sub> +2V			V		
Input Bias Current of Enable Pin				5		pA		
$I_{QSD}$		Amplifier Disabled		6	15	μΑ		
POWER SUPPLY								
Specified Voltage Range	Vs		4		12	V		
Operating Voltage Range	Vs			3.5 to 13.2		V		
Quiescent Current (per amplifier)	IQ			4.3	6.5	mA		
Over Temperature					6.5	mA		
TEMPERATURE RANGE								
Specified Range			-40		+125	°C		
Operating Range			-55		+125	°C		
Storage Range			-55		+150	°C		
Thermal Resistance	$\theta_{JA}$							
MSOP-8, SO-8				150		°C/W		
TSSOP-14				100		°C/W		
DFN-8				46		°C/W		



#### **TYPICAL CHARACTERISTICS**

At  $T_A$  = +25°C,  $V_S$  = ±6V,  $R_L$  = 10k $\Omega$  connected to  $V_S/2$ , and  $V_{OUT}$  =  $V_S/2$ , unless otherwise noted.

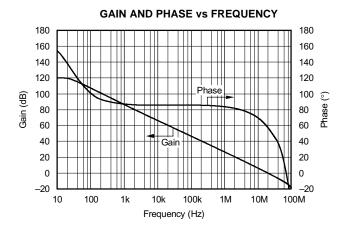


Figure 1.

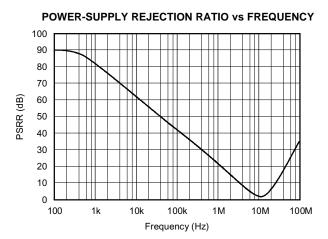


Figure 3.

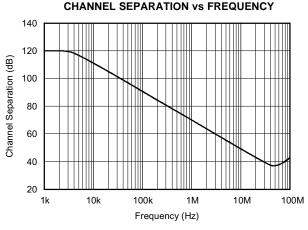


Figure 5.

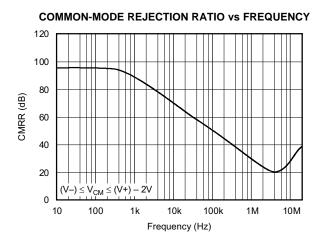


Figure 2.

#### **MAXIMUM OUTPUT VOLTAGE vs FREQUENCY**

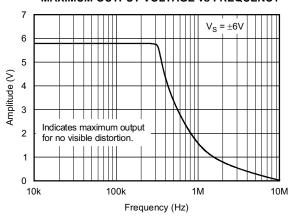


Figure 4.

### INPUT VOLTAGE NOISE SPECTRAL DENSITY

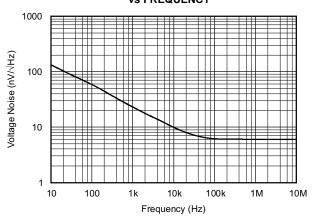


Figure 6.



At  $T_A$  = +25°C,  $V_S$  = ±6V,  $R_L$  = 10k $\Omega$  connected to  $V_S/2$ , and  $V_{OUT}$  =  $V_S/2$ , unless otherwise noted.

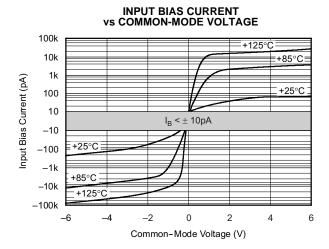


Figure 7.

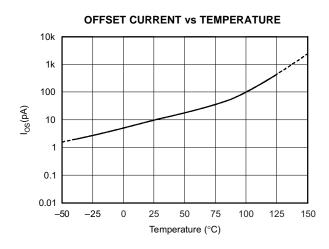


Figure 8.

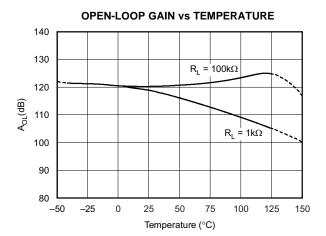


Figure 9.

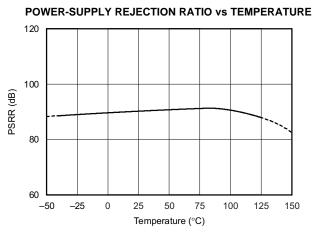
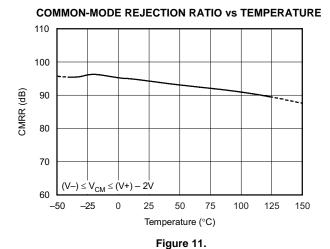


Figure 10.

**QUIESCENT CURRENT vs TEMPERATURE** 



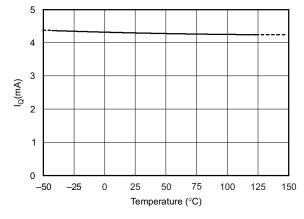


Figure 12.



At  $T_A$  = +25°C,  $V_S$  = ±6V,  $R_L$  = 10k $\Omega$  connected to  $V_S/2$ , and  $V_{OUT}$  =  $V_S/2$ , unless otherwise noted.

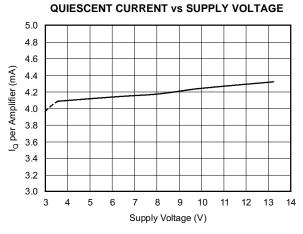


Figure 13.

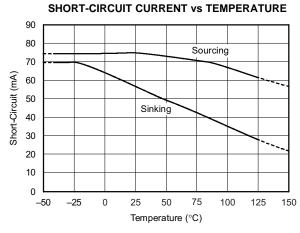


Figure 14.

#### SHORT-CIRCUIT CURRENT vs SUPPLY VOLTAGE

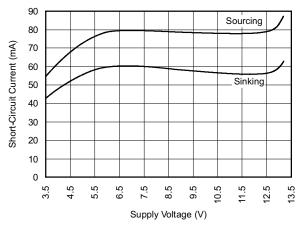


Figure 15.

# OUTPUT VOLTAGE SWING vs OUTPUT CURRENT

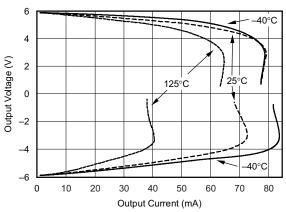


Figure 16.

#### TOTAL HARMONIC DISTORTION + NOISE vs FREQUENCY

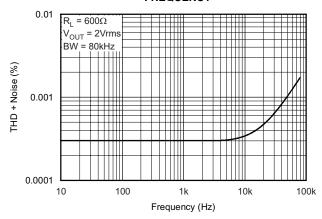


Figure 17.

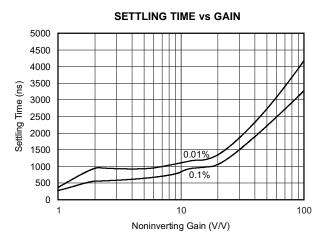


Figure 18.



At  $T_A$  = +25°C,  $V_S$  = ±6V,  $R_L$  = 10k $\Omega$  connected to  $V_S/2$ , and  $V_{OUT}$  =  $V_S/2$ , unless otherwise noted.

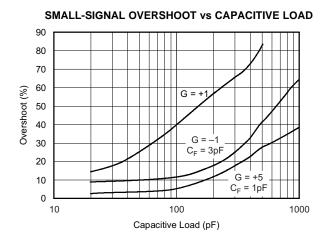


Figure 19.

# OFFSET VOLTAGE DRIFT PRODUCTION DISTRIBUTION (0°C TO +85°C)

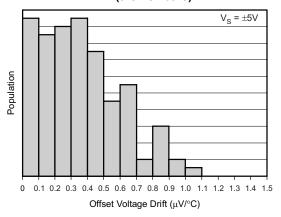


Figure 21.

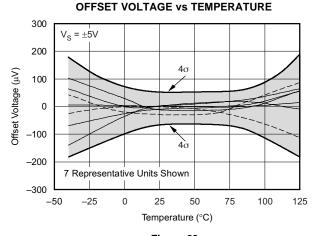


Figure 23.

#### OFFSET VOLTAGE PRODUCTION DISTRIBUTION

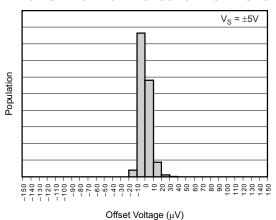


Figure 20.

# OFFSET VOLTAGE DRIFT PRODUCTION DISTRIBUTION (-40°C TO +125°C)

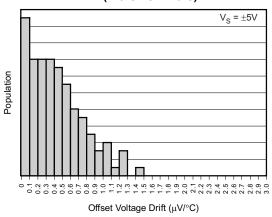


Figure 22.

#### SMALL-SIGNAL STEP RESPONSE

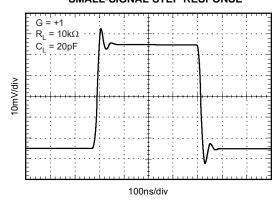
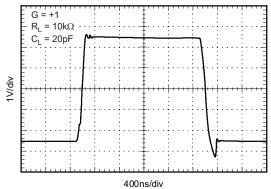


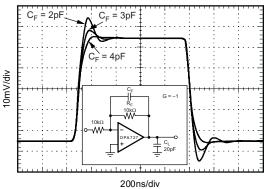
Figure 24.



At  $T_A$  = +25°C,  $V_S$  = ±6V,  $R_L$  = 10k $\Omega$  connected to  $V_S/2$ , and  $V_{OUT}$  =  $V_S/2$ , unless otherwise noted.

#### LARGE-SIGNAL STEP RESPONSE





SMALL-SIGNAL STEP RESPONSE

25. Figure 26.

Figure 25.

#### LARGE-SIGNAL STEP RESPONSE

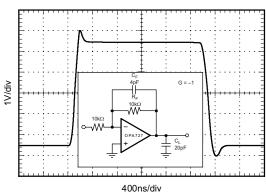


Figure 27.



#### APPLICATIONS INFORMATION

The OPA727 and OPA728 family of op amps use *e-trim*, an adjustment to offset voltage and temperature drift made during the final steps of manufacturing after the plastic molding is completed. This compensates for performance shifts that can occur during the molding process. Through e-trim, the OPA727 and OPA728 deliver excellent offset voltage (150 $\mu$ V max) and extremely low offset voltage drift (1.5 $\mu$ V/°C). Additionally, these 20MHz CMOS op amps have a fast slew rate, low noise, and excellent PSRR, CMRR, and A<sub>OL</sub>. They can operate on typically 4.3mA quiescent current from a single (or split) supply in the range of 4V to 12V ( $\pm$ 2V to  $\pm$ 6V), making them highly versatile and easy to use. They are stable in a unity-gain configuration.

Power-supply pins should be bypassed with 1nF ceramic capacitors in parallel with  $1\mu\text{F}$  tantalum capacitors.

#### **OPERATING VOLTAGE**

OPA727 series op amps are specified from 4V to 12V supplies over a temperature range of  $-40^{\circ}$ C to +125°C. They will operate well in  $\pm 5$ V or +5V to +12V power-supply systems. Parameters that vary significantly with operating voltage or temperature are shown in the Typical Characteristics.

#### **ENABLE/SHUTDOWN**

OPA727 series op amps require approximately 4.3mA quiescent current. The enable/shutdown feature of the OPA728 allows the op amp to be shut off to reduce this current to approximately  $6\mu\text{A}$ .

The enable/shutdown input is referenced to the Enable Reference Pin, REF (see *Pin Configurations*). This pin can be connected to logic ground in dual-supply op amp configurations to avoid level-shifting the enable logic signal, as shown in Figure 28.

The Enable Reference Pin voltage,  $V_{REF}$ , must not exceed (V+) – 2V. It may be set as low as V–. The amplifier is enabled when the Enable Pin voltage is greater than  $V_{REF}$  + 2V. The amplifier is disabled (shutdown) if the Enable Pin voltage is less than  $V_{REF}$  + 0.8V. The Enable Pin is connected to internal pull-up circuitry and will enable the device if left unconnected.

#### **COMMON-MODE VOLTAGE RANGE**

The input common-mode voltage range of the OPA727 and OPA728 series extends from V- to (V+)-2.5V.

Common-mode rejection is excellent throughout the input voltage range from V- to (V+) – 3V. CMRR decreases somewhat as the common-mode voltage extends to (V+) – 2.5V, but remains very good and is tested throughout this range. See the *Electrical Characteristics* table for details.

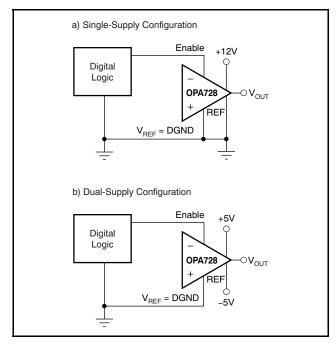


Figure 28. Enable Reference Pin Connection for Single- and Dual-Supply Configurations

#### INPUT OVER-VOLTAGE PROTECTION

Device inputs are protected by ESD diodes that will conduct if the input voltages exceed the power supplies by more than approximately 300mV. Momentary voltages greater than 300mV beyond the power supply can be tolerated if the current is limited to 10mA. This is easily accomplished with an input resistor in series with the op amp, as shown in Figure 29. The OPA727 series features no phase inversion when the inputs extend beyond supplies, if the input is current limited.

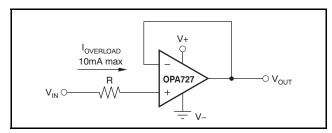


Figure 29. Input Current Protection for Voltages Exceeding the Supply Voltage



#### **RAIL-TO-RAIL OUTPUT**

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. This output stage is capable of driving heavy loads connected to any point between V+ and V–. For light resistive loads (>100k $\Omega$ ), the output voltage can swing to 150mV from the supply rail, while still maintaining excellent linearity (A<sub>OL</sub> > 110dB). With 1k $\Omega$  resistive loads, the output is specified to swing to within 250mV from the supply rails with excellent linearity (see the Typical Characteristics curve, Output Voltage Swing vs Output Current).

#### **CAPACITIVE LOAD AND STABILITY**

Capacitive load drive is dependent upon gain and the overshoot requirements of the application. Increasing the gain enhances the ability of the amplifier to drive greater capacitive loads (see the Typical Characteristics curve, Small-Signal Overshoot vs Capacitive Load).

One method of improving capacitive load drive in the unity-gain configuration is to insert a  $10\Omega$  to  $20\Omega$  resistor inside the feedback loop, as shown in Figure 30. This reduces ringing with large capacitive loads while maintaining DC accuracy.

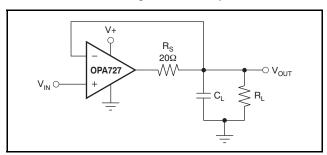


Figure 30. Series Resistor in Unity-Gain Buffer Configuration Improves Capacitive Load Drive

#### **DRIVING FAST 16-BIT ADCs**

The OPA727 series is optimized for driving fast 16-bit ADCs such as the ADS8342. The OPA727 op amps buffer the converter input capacitance and resulting charge injection, while providing signal gain. Figure 31 shows the OPA727 in a single-ended method of interfacing to the ADS8342 16-bit, 250kSPS, 4-channel ADC with an input range of ±2.5V. The OPA727 has demonstrated excellent settling time to the 16-bit level within the 600ns acquisition time of the ADS8342. The RC filter, shown in Figure 31, has been carefully tuned for best noise and settling performance. It may need to be adjusted for different op amp configurations. Refer to the ADS8342 data sheet (available for download at www.ti.com) for additional information on this product.

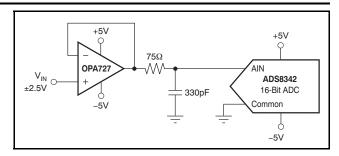


Figure 31. OPA727 Driving an ADC

#### TRANSIMPEDANCE AMPLIFIER

Wide bandwidth, low input bias current, and low input voltage and current noise make the OPA727 an ideal wideband photodiode transimpedance amplifier. Low-voltage noise is important because photodiode capacitance causes the effective noise gain of the circuit to increase at high frequency.

The key elements to a transimpedance design, as shown in Figure 32, are the expected diode capacitance ( $C_D$ ), which should include the parasitic input common-mode and differential-mode input capacitance (4pF + 5pF for the OPA727); the desired transimpedance gain ( $R_F$ ); and the GBW for the OPA727 (20MHz). With these three variables set, the feedback capacitor value ( $C_F$ ) can be set to control the frequency response.  $C_F$  includes the stray capacitance of  $R_F$ , which is 0.2pF for a typical surface-mount resistor.

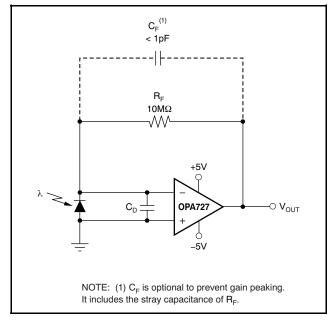


Figure 32. Dual-Supply Transimpedance Amplifier



To achieve a maximally-flat, 2nd-order Butterworth frequency response, the feedback pole should be set to:

$$\frac{1}{2\pi R_F C_F} = \sqrt{\frac{GBW}{4\pi R_F C_D}}$$
 (1)

Bandwidth is calculated by:

$$f_{-3dB} = \sqrt{\frac{GBW}{2\pi R_F C_D}} Hz$$
 (2)

For even higher transimpedance bandwidth, the high-speed CMOS OPA380 (90MHz GBW), OPA354 (100MHz GBW), OPA300 (180MHz GBW), OPA355 (200MHz GBW), or OPA656, OPA657 (400MHz GBW) may be used.

For single-supply applications, the +IN input can be biased with a positive dc voltage to allow the output to reach true zero when the photodiode is not exposed to any light, and respond without the added delay that results from coming out of the negative rail; this is shown in Figure 33. This bias voltage also appears across the photodiode, providing a reverse bias for faster operation.

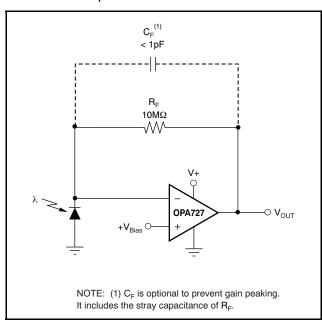


Figure 33. Single-Supply Transimpedance Amplifier

For additional information, refer to Application Bulletin (SBOA055), Compensate Transimpedance Amplifiers Intuitively, available for download at www.ti.com.

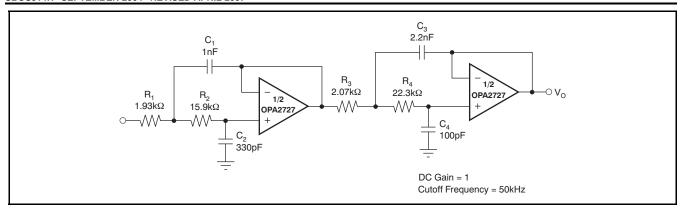
# OPTIMIZING THE TRANSIMPEDANCE CIRCUIT

To achieve the best performance, components should be selected according to the following guidelines:

- For lowest noise, select R<sub>F</sub> to create the total required gain. Using a lower value for R<sub>F</sub> and adding gain after the transimpedance amplifier generally produces poorer noise performance. The noise produced by R<sub>F</sub> increases with the square-root of R<sub>F</sub>, whereas the signal increases linearly. Therefore, signal-to-noise ratio is improved when all the required gain is placed in the transimpedance stage.
- 2. Minimize photodiode capacitance and stray capacitance at the summing junction (inverting input). This capacitance causes the voltage noise of the op amp to be amplified (increasing amplification at high frequency). Using a low-noise voltage source to reverse-bias a photodiode can significantly reduce its capacitance. Smaller photodiodes have lower capacitance. Use optics to concentrate light on a small photodiode.
- 3. Noise increases with increased bandwidth. Limit the circuit bandwidth to only that required. Use a capacitor across the  $R_{\text{F}}$  to limit bandwidth, even if not required for stability.
- 4. Circuit board leakage can degrade the performance of an otherwise well-designed amplifier. Clean the circuit board carefully. A circuit board guard trace that encircles the summing junction and is driven at the same voltage can help control leakage.

For additional information, refer to the Application Bulletins *Noise Analysis of FET Transimpedance Amplifiers* (SBOA060), and *Noise Analysis for High-Speed Op Amps* (SBOA066), available for download at the TI web site.





Note: FilterPro is a low-pass filter design program available for download at no cost from Tl's web site (www.ti.com). The program can be used to determine component values for other cutoff frequencies or filter types.

Figure 34. Four-Pole Butterworth Sallen-Key Low-Pass Filter

#### **DFN PACKAGE**

The OPA727 series uses the DFN-8 (also known as SON), which is a QFN package with lead contacts on only two sides of the bottom of the package. This leadless, near-chip-scale package maximizes board space and enhances thermal and electrical characteristics through an exposed pad.

DFN packages are physically small, have a smaller routing area, improved thermal performance, and improved electrical parasitics, with a pinout scheme that is consistent with other commonly-used packages, such as SO and MSOP. Additionally, the absence of external leads eliminates bent-lead issues.

The DFN package can be easily mounted using standard printed circuit board (PCB) assembly techniques. See Application Note, *QFN/SON PCB Attachment* (SLUA271) and Application Report, *Quad Flatpack No-Lead Logic Packages* (SCBA017), both available for download at www.ti.com.

The exposed leadframe die pad on the bottom of the package should be connected to V-.

#### **LAYOUT GUIDELINES**

The leadframe die pad should be soldered to a thermal pad on the PCB. A mechanical data sheet showing an example layout is attached at the end of this data sheet. Refinements to this layout may be required based on assembly process requirements. Mechanical drawings located at the end of this data sheet list the physical dimensions for the package and pad. The five holes in the landing pattern are optional, and are intended for use with thermal vias that connect the leadframe die pad to the heatsink area on the PCB.

Soldering the exposed pad significantly improves board-level reliability during temperature cycling, key push, package shear, and similar board-level tests. Even with applications that have low-power dissipation, the exposed pad must be soldered to the PCB to provide structural integrity and long-term reliability.

20-Feb-2012

#### **PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
OPA2727AID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDRBR	ACTIVE	SON	DRB	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDRBRG4	ACTIVE	SON	DRB	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDRBT	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDRBTG4	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA2727AIDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
OPA4727AIPW	ACTIVE	TSSOP	PW	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA4727AIPWG4	ACTIVE	TSSOP	PW	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA4727AIPWR	ACTIVE	TSSOP	PW	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA4727AIPWRG4	ACTIVE	TSSOP	PW	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA727AIDGKR	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA727AIDGKRG4	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA727AIDGKT	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA727AIDGKTG4	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA727AIDRBT	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	



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### PACKAGE OPTION ADDENDUM

20-Feb-2012

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
OPA727AIDRBTG4	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA728AIDGKT	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAUAG	Level-2-260C-1 YEAR	
OPA728AIDGKTG4	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAUAG	Level-2-260C-1 YEAR	
OPA728AIDRBT	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA728AIDRBTG4	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL. Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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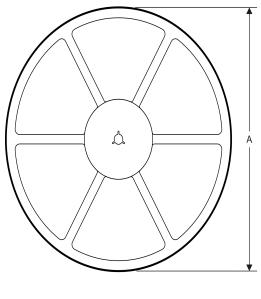
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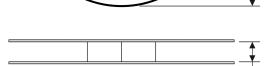
### PACKAGE MATERIALS INFORMATION

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#### TAPE AND REEL INFORMATION

#### **REEL DIMENSIONS**





#### **TAPE DIMENSIONS**



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### TAPE AND REEL INFORMATION

#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2727AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA2727AIDRBR	SON	DRB	8	2500	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA2727AIDRBT	SON	DRB	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA4727AIPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
OPA727AIDGKR	MSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA727AIDGKT	MSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA727AIDRBT	SON	DRB	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA728AIDGKT	MSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA728AIDRBT	SON	DRB	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

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\*All dimensions are nominal

7 til dillionorono dro morninar							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2727AIDR	SOIC	D	8	2500	346.0	346.0	29.0
OPA2727AIDRBR	SON	DRB	8	2500	346.0	346.0	29.0
OPA2727AIDRBT	SON	DRB	8	250	210.0	185.0	35.0
OPA4727AIPWR	TSSOP	PW	14	2000	346.0	346.0	29.0
OPA727AIDGKR	MSOP	DGK	8	2500	346.0	346.0	29.0
OPA727AIDGKT	MSOP	DGK	8	250	210.0	185.0	35.0
OPA727AIDRBT	SON	DRB	8	250	210.0	185.0	35.0
OPA728AIDGKT	MSOP	DGK	8	250	210.0	185.0	35.0
OPA728AIDRBT	SON	DRB	8	250	210.0	185.0	35.0

# DGK (S-PDSO-G8)

## PLASTIC SMALL-OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
- E. Falls within JEDEC MO-187 variation AA, except interlead flash.



DRB (S-PVSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Small Outline No-Lead (SON) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.



### DRB (S-PVSON-N8)

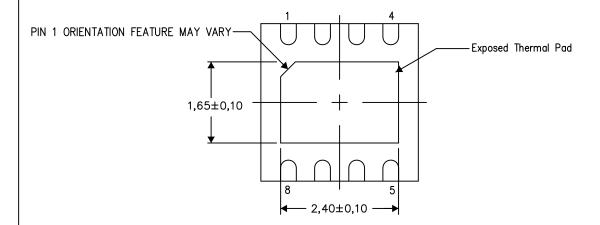
### PLASTIC SMALL OUTLINE NO-LEAD

#### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

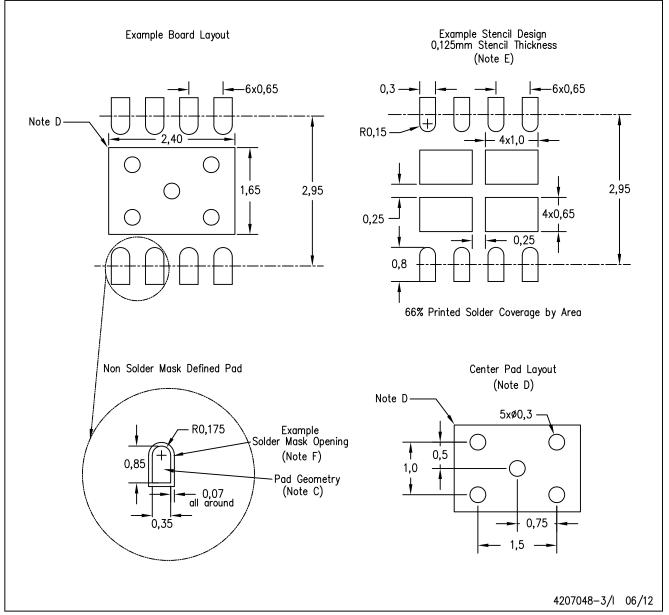
4206340-3/M 06/12

NOTE: All linear dimensions are in millimeters



### DRB (S-PVSON-N8)

### PLASTIC SMALL OUTLINE NO-LEAD



- S: A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="https://www.ti.com">http://www.ti.com</a>.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
  - F. Customers should contact their board fabrication site for solder mask tolerances.



PW (R-PDSO-G14)

### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
  - Sody length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
- E. Falls within JEDEC MO-153



# PW (R-PDSO-G14)

### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



### D (R-PDSO-G8)

#### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



# D (R-PDSO-G8)

### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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