# LMC7101,LMC7101Q

LMC7101/LMC7101Q Tiny Low Power Operational Amplifier with Rail-to-Rail Input and Output



Literature Number: SNOS719E



### LMC7101/LMC7101Q

# Tiny Low Power Operational Amplifier with Rail-to-Rail Input and Output

### **General Description**

The LMC7101 is a high performance CMOS operational amplifier available in the space saving 5-Pin SOT23 Tiny package. This makes the LMC7101 ideal for space and weight critical designs. The performance is similar to a single amplifier of the LMC6482/LMC6484 type, with rail-to-rail input and output, high open loop gain, low distortion, and low supply currents.

The main benefits of the Tiny package are most apparent in small portable electronic devices, such as mobile phones, pagers, notebook computers, personal digital assistants, and PCMCIA cards. The tiny amplifiers can be placed on a board where they are needed, simplifying board layout.

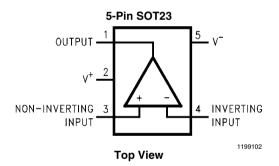
#### **Features**

- Tiny 5-Pin SOT23 package saves space—typical circuit layouts take half the space of 8-Pin SOIC designs
- Guaranteed specs at 2.7V, 3V, 5V, 15V supplies
- Typical supply current 0.5 mA at 5V
- Typical total harmonic distortion of 0.01% at 5V
- 1.0 MHz gain-bandwidth
- Similar to popular LMC6482/LMC6484
- Rail-to-rail input and output
- Temperature Range -40°C to 125°C (LMC7101Q)

### **Applications**

- Mobile communications
- Notebooks and PDAs
- Battery powered products
- Sensor interface
- Automotive applications (LMC7101Q)

### **Connection Diagram**



# **Ordering Information**

Package	Part Number	Package Marking	Transport Media	NSC Drawing	Features
	LMC7101AIM5	A00A	1k Units on Tape and Reel		
	LMC7101AIM5X	AUUA	3k Units Tape and Reel		
5-Pin SOT23	LMC7101BIM5	A00B	1k Units on Tape and Reel	MF05A	
5-111 50123	LMC7101BIM5X	AUUD	3k Units Tape and Reel	IVIFUSA	
	LMC7101QM5	AT6A	1k Units on Tape and Reel		-40°C to 125°C
	LMC7101QM5X	ATOA	3k Units Tape and Reel		Operating range

<sup>\*</sup> The LMC7101Q incorporates enhanced manufacturing and support processes for the automotive market, including defect detection methodologies.

### **Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Tolerance (Note 2)

Human Body Model 1000V Machine Model 200V Charged Device Model 1000V

Difference Input Voltage  $\pm$ Supply Voltage Voltage at Input/Output Pin  $(V^+) + 0.3V, (V^-) - 0.3V$ 

Supply Voltage  $(V^+ - V^-)$  16V Current at Input Pin  $\pm 5$  mA

Current at Output Pin (Note 3) ±35 mA

Current at Power Supply Pin 35 mA

Lead Temp. (Soldering, 10 sec.) 260°C Storage Temperature Range -65°C to +150°C Junction Temperature (Note 4) 150°C

# **Recommended Operating**

**Conditions** (Note 1)

Supply Voltage  $2.7V \le V^{+} \le 15.5V$ 

Temperature Range

LMC7101AI, LMC7101BI -40°C to 85°C LMC7101Q -40°C to 125°C

Thermal Resistance (θ<sub>.IA</sub>)

5-Pin SOT23 325°C/W

### 2.7V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25$ °C,  $V^+ = 2.7V$ ,  $V^- = 0V$ ,  $V_{CM} = V_O = V^+/2$  and  $R_L > 1$  M $\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LMC7101AI Limit (Note 6)	LMC7101BI Limit (Note 6)	LMC7101Q Limit (Notes 6, 10)	Units
V <sub>os</sub>	Input Offset Voltage Average Drift	V+ = 2.7V	0.11	6	9	9	mV max
TCV <sub>OS</sub>	Input Offset Voltage		1				μV/°C
I <sub>B</sub>	Input Bias Current		1.0	64	64	1000	pA max
I <sub>os</sub>	Input Offset Current		0.5	32	32	2000	pA max
R <sub>IN</sub>	Input Resistance		>1				Tera Ω
CMRR	Common-Mode Rejection Ratio	$0V \le V_{CM} \le 2.7V$ $V^{+} = 2.7V$	70	55	50	50	dB min
V <sub>CM</sub>	Input Common Mode Voltage	F- :: OMDD > 50 -ID	0.0	0.0	0.0	0.0	V min
	Range	For CMRR ≥ 50 dB	3.0	2.7	2.7	2.7	V max
PSRR	Power Supply Rejection Ratio	$V^{+} = 1.35V$ to $1.65V$ $V^{-} = -1.35V$ to $-1.65V$ $V_{CM} = 0$	60	50	45	45	dB min
C <sub>IN</sub>	Common-Mode Input Capacitance		3				pF
		$R_1 = 2 k\Omega$	2.45	2.15	2.15	2.15	V min
$V_{O}$	Output Swing	n <sub>L</sub> – 2 kg	0.25	0.5	0.5	0.5	V max
<b>v</b> O	Output Swing	$R_1 = 10 \text{ k}\Omega$	2.68	2.64	2.64	2.64	V min
		11[ = 10 K22	0.025	0.06	0.06	0.06	V max
I <sub>S</sub>	Supply Current		0.5	0.81 <b>0.95</b>	0.81 <b>0.95</b>	0.81 <b>0.95</b>	mA max
SR	Slew Rate (Note 8)		0.7				V/µs
GBW	Gain-Bandwidth Product		0.6				MHz

# **3V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J$  = 25°C,  $V^+$  = 3V,  $V^-$  = 0V,  $V_{CM}$  = 1.5V,  $V_O$  = V+/2 and  $R_L$  = 1 M $\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LMC7101AI Limit (Note 6)	LMC7101BI Limit (Note 6)	LMC7101Q Limit (Notes 6, 10)	Units
V <sub>OS</sub>	Input Offset Voltage		0.11	4 <b>6</b>	7 <b>9</b>	7	mV max
TCV <sub>OS</sub>	Input Offset Voltage Average Drift		1				μV/°C
I <sub>B</sub>	Input Current		1.0	64	64	1000	pA max
I <sub>os</sub>	Input Offset Current		0.5	32	32	2000	pA max
R <sub>IN</sub>	Input Resistance		>1				Tera Ω
CMRR	Common-Mode Rejection Ratio	$0V \le V_{CM} \le 3V$ $V^{+} = 3V$	74	64	60	60	db min
V	Input Common-Mode Voltage Range	E 0MPD > 50 ID	0.0	0.0	0.0	0.0	V min
V <sub>CM</sub>		For CMRR ≥ 50 dB	3.3	3.0	3.0	3.0	V max
PSRR	Power Supply Rejection Ratio	$V^{+} = 1.5V \text{ to } 7.5V$ $V^{-} = -1.5V \text{ to } -7.5V$ $V_{O} = V_{CM} = 0$	80	68	60	60	dBmin
C <sub>IN</sub>	Common-Mode Input Capacitance		3				pF
		D 010	2.8	2.6	2.6	2.6	V min
Vo	Output Swing	$R_L = 2 k\Omega$	0.2	0.4	0.4	0.4	V max
<b>v</b> O	Output Swing	$R_1 = 600\Omega$	2.7	2.5	2.5	2.5	V min
		11 - 00012	0.37	0.6	0.6	0.6	V max
I <sub>s</sub>	Supply Current		0.5	0.81 <b>0.95</b>	0.81 <b>0.95</b>	0.81 <b>0.95</b>	mA max

### **5V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 5V$ ,  $V^- = 0V$ ,  $V_{CM} = 1.5V$ ,  $V_O = V^+/2$  and  $R_L = 1$  M $\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions		Typ (Note 5)	LMC7101AI Limit (Note 6)	LMC7101BI Limit (Note 6)	LMC7101Q Limit (Notes 6, 10)	Units
V <sub>OS</sub>	Input Offset Voltage	V+ = 5V		0.11	3 <b>5</b>	7 <b>9</b>	7 <b>9</b>	mV max
TCV <sub>os</sub>	Input Offset Voltage Average Drift			1.0				μV/°C
I <sub>B</sub>	Input Current			1	64	64	1000	pA max
I <sub>os</sub>	Input Offset Current			0.5	32	32	2000	pA max
R <sub>IN</sub>	Input Resistance			>1				Tera Ω
CMRR	Common-Mode Rejection Ratio	$0V \le V_{CM} \le 5V$ LMC7101Q @ 125°C $0.2V \le V_{CM} \le 4.8V$		82	65 <b>60</b>	60 <b>55</b>	60 <b>55</b>	db min
+PSRR	Positive Power Supply Rejection Ratio	$V^{+} = 5V \text{ to } 15V$ $V^{-} = 0V, V_{O} = 0$		82	70 <b>65</b>	65 <b>62</b>	65 <b>62</b>	dB min
-PSRR	Negative Power Supply Rejection Ratio	$V^{-} = -5V \text{ to } -$ $V^{+} = 0V, V_{O} =$		82	70 <b>65</b>	65 <b>62</b>	65 <b>62</b>	dB min
V <sub>CM</sub>	Input Common-Mode Voltage	For CMRR ≥ 50 dB		-0.3	-0.20 <b>0.00</b>	-0.20 <b>0.00</b>	-0.2 <b>0.2</b>	V min
V CM	Range			5.3	5.20 <b>5.00</b>	5.20 <b>5.00</b>	5.2 <b>4.8</b>	V max
C <sub>IN</sub>	Common-Mode Input Capacitance			3				pF
		$R_L = 2 k\Omega$		4.9	4.7 <b>4.6</b>	4.7 <b>4.6</b>	4.7 <b>4.54</b>	V min
.,	0.4.40.4			0.1	0.18 <b>0.24</b>	0.18 <b>0.24</b>	0.18 <b>0.28</b>	V max
V <sub>o</sub>	Output Swing	R <sub>L</sub> = 600Ω		4.7	4.5 <b>4.24</b>	4.5 <b>4.24</b>	4.5 <b>4.28</b>	V min
				0.3	0.5 <b>0.65</b>	0.5 <b>0.65</b>	0.5 <b>0.8</b>	V max
	Output Chart Circuit Cours	V <sub>O</sub> = 0V 24	Sourcing	24	16 <b>11</b>	16 <b>11</b>	16 <b>9</b>	mA min
I <sub>SC</sub>	Output Short Circuit Current	V <sub>O</sub> = 5V	Sinking	19	11 <b>7.</b> 5	11 <b>7.</b> 5	11 <b>5.8</b>	mA min
I <sub>S</sub>	Supply Current		•	0.5	0.85 <b>1.0</b>	0.85 <b>1.0</b>	0.85 <b>1.0</b>	mA max

### **5V AC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 5V$ ,  $V^- = 0V$ ,  $V_{CM} = 1.5V$ ,  $V_O = V^+/2$  and  $R_L = 1$  M $\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LMC7101AI Limit (Note 6)	LMC7101BI Limit (Note 6)	Units
THD	Total Harmonic Distortion	$f = 10 \text{ kHz}, A_V = -2$ $R_L = 10 \text{ k}\Omega, V_O = 4.0 V_{PP}$	0.01			%
SR	Slew Rate		1.0			V/µs
GBW	Gain Bandwidth Product		1.0			MHz

# **15V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 15V$ ,  $V^- = 0V$ ,  $V_{CM} = 1.5V$ ,  $V_O = V^+/2$  and  $R_L = 1~M\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions		Typ (Note 5)	LMC7101AI Limit (Note 6)	LMC7101BI Limit (Note 6)	LMC7101Q Limit (Notes 6, 10)	Units
Vos	Input Offset Voltage			0.11				mV max
TCV <sub>OS</sub>	Input Offset Voltage Average Drift			1.0				μV/°C
I <sub>B</sub>	Input Current			1.0	64	64	1000	pA max
I <sub>os</sub>	Input Offset Current			0.5	32	32	2000	pA max
R <sub>IN</sub>	Input Resistance			>1				Tera Ω
CMRR	Common-Mode Rejection Ratio	0V ≤ V <sub>CM</sub> ≤ 15V LMC7101Q @°125C 0.2V ≤ V <sub>CM</sub> ≤ 14.8V		82	70 <b>65</b>	65 <b>60</b>	65 <b>60</b>	dB min
+PSRR	Positive Power Supply Rejection Ratio	$V^{+} = 5V \text{ to } 15V$ $V^{-} = 0V, V_{O} = 1.5V$		82	70 <b>65</b>	65 <b>62</b>	65 <b>62</b>	dB min
-PSRR	Negative Power Supply Rejection Ratio	$V^{-} = -5V \text{ to } -15V$ $V^{+} = 0V, V_{O} = -1.5V$		82	70 <b>65</b>	65 <b>62</b>	65 <b>62</b>	dB min
$V_{CM}$	Input Common-Mode Voltage	V+ = 5V For CMRR ≥ 50 dB		-0.3	-0.20 <b>0.00</b>	-0.20 <b>0.00</b>	-0.2 <b>0.2</b>	V min
OW	Range			15.3	15.20 <b>15.00</b>	15.20 <b>15.00</b>	15.2 <b>14.8</b>	V max
	Large Signal Voltage Gain (Note 7)	$R_L = 2 k\Omega$	Sourcing	340	80 <b>40</b>	80 <b>40</b>	80 <b>30</b>	V/mV
$A_V$		L		24	15 <b>10</b>	15 <b>10</b>	15 <b>4</b>	
		$R_L = 600\Omega$	Sourcing Sinking	300 15	34 6	34 6	34 6	V/mV
C <sub>IN</sub>	Input Capacitance			3				pF
		$V^{+} = 15V$ $R_L = 2 k\Omega$		14.7	14.4 <b>14.2</b>	14.4 <b>14.2</b>	14.4 <b>14.2</b>	V min
V <sub>o</sub>	Output Swing			0.16	0.32 <b>0.45</b>	0.32 <b>0.45</b>	0.32 <b>0.45</b>	V max
<b>v</b> 0	Output Swing	$V^{+} = 15V$ $R_{L} = 600\Omega$		14.1	13.4 <b>13.0</b>	13.4 <b>13.0</b>	13.4 <b>12.85</b>	V min
				0.5	1.0 <b>1.3</b>	1.0 <b>1.3</b>	1.0 <b>1.5</b>	V max
l	Output Short Circuit Current	V <sub>O</sub> = 0V	Sourcing	50	30 <b>20</b>	30 <b>20</b>	30 <b>20</b>	mA min
I <sub>SC</sub>	(Note 9)	V <sub>O</sub> = 12V	Sinking	50	30 <b>20</b>	30 <b>20</b>	30 <b>20</b>	
I <sub>S</sub>	Supply Current			0.8	1.50 <b>1.71</b>	1.50 <b>1.71</b>	1.50 <b>1.75</b>	mA max

### 15V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 15V$ ,  $V^- = 0V$ ,  $V_{CM} = 1.5V$ ,  $V_O = V^+/2$  and  $R_L = 1 \text{ M}\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LMC7101AI Limit (Note 6)	LMC7101BI Limit (Note 6)	LMC7101Q Limit (Notes 6, 10)	Units
SR	Slew Rate (Note 8)	V+ = 15V	1.1	0.5 <b>0.4</b>	0.5 <b>0.4</b>	0.5 <b>0.4</b>	V/µs min
GBW	Gain-Bandwidth Product	V+ = 15V	1.1				MHz
$\overline{\phi_{m}}$	Phase Margin		45				deg
G <sub>m</sub>	Gain Margin		10				dB
e <sub>n</sub>	Input-Referred Voltage Noise	f = 1 kHz, V <sub>CM</sub> = 1V	37				$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
i <sub>n</sub>	Input-Referred Current Noise	f = 1 kHz	1.5				$\frac{fA}{\sqrt{Hz}}$
THD	Total Harmonic Distortion	f = 10 kHz, $A_V = -2$ $R_L = 10 kΩ$ , $V_O = 8.5 V_{PP}$	0.01				%

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Human Body Model is  $1.5 \text{ k}\Omega$  in series with 100 pF.

Note 3: Applies to both single-supply and split-supply operation. Continuous short operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature at 150°C.

Note 4: The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$  and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)}, -T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

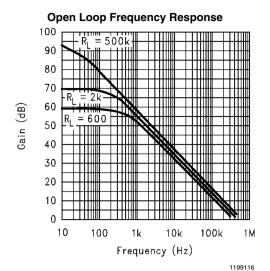
Note 7: V+ = 15V, V<sub>CM</sub> = 1.5V and R<sub>L</sub> connect to 7.5V. For sourcing tests, 7.5V  $\leq$  V<sub>O</sub>  $\leq$  12.5V. For sinking tests, 2.5V  $\leq$  V<sub>O</sub>  $\leq$  7.5V.

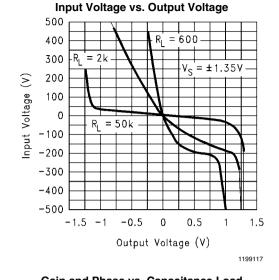
Note 8: V+ = 15V. Connected as a voltage follower with a 10V step input. Number specified is the slower of the positive and negative slew rates.  $R_L = 100 \text{ k}\Omega$  connected to 7.5V. Amp excited with 1 kHz to produce  $V_Q = 10 \text{ V}_{PP}$ .

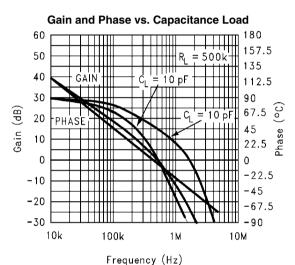
Note 9: Do not short circuit output to V+ when V+ is greater than 12V or reliability will be adversely affected.

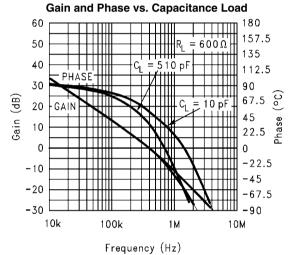
Note 10: When operated at temperature between -40°C and 85°C, the LMC7101Q will meet LMC7101BI specifications.

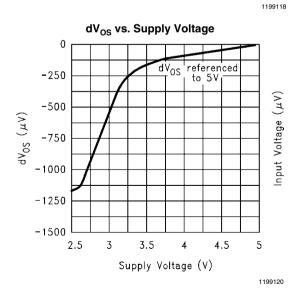
# 2.7V Typical Performance Characteristics V+=2.7V, V-=0V, T<sub>A</sub>=25°C, unless otherwise specified.

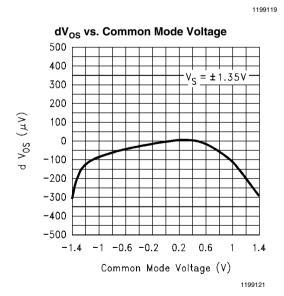


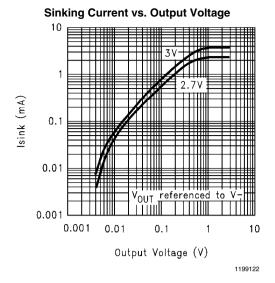


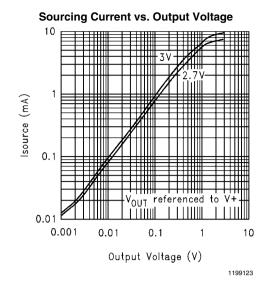




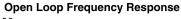


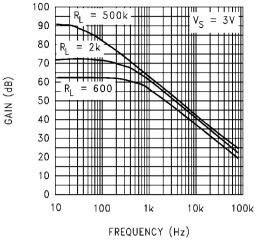






# **3V Typical Performance Characteristics** $V^+ = 3V$ , $V^- = 0V$ , $T_A = 25$ °C, unless otherwise specified.





### 500 = 600 400 300 VOLTAGE $(\mu V)$ 200 100 50k -100 -200 -300 -400 -500

-1.0

-0.5

Input Voltage vs. Output Voltage

0.0

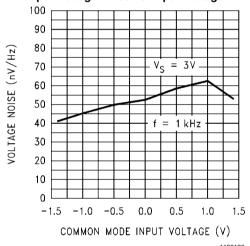
OUTPUT VOLTAGE (V)

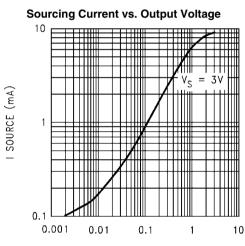
0.5

1.0

1199125

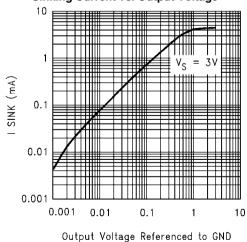
#### Input Voltage Noise vs. Input Voltage



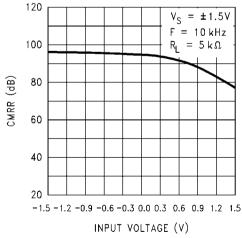


Output Voltage Referenced to  $V_{\rm S}$ 

### Sinking Current vs. Output Voltage



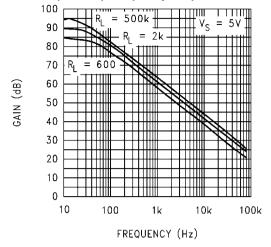
CMRR vs. Input Voltage



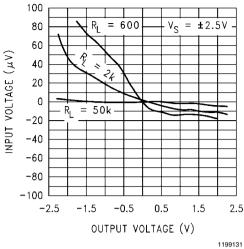
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# **5V Typical Performance Characteristics** $V^+ = 5V$ , $V^- = 0V$ , $T_A = 25$ °C, unless otherwise specified.

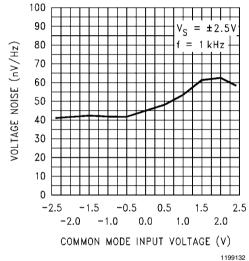




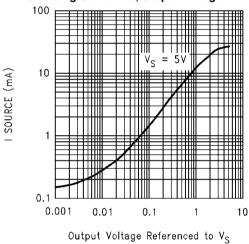
### Input Voltage vs. Output Voltage



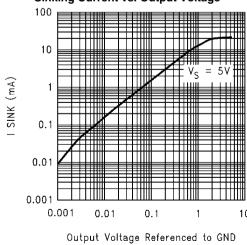
#### Input Voltage Noise vs. Input Voltage



#### Sourcing Current vs, Output Voltage

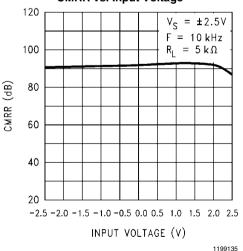


#### Sinking Current vs. Output Voltage



#### CMRR vs. Input Voltage

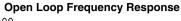
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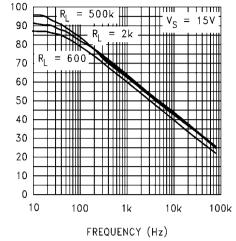


# **15V Typical Performance Characteristics** $V^+ = +15V$ , $V^- = 0V$ , $T_A = 25$ °C, unless otherwise specified.

100

80





GAIN (dB)

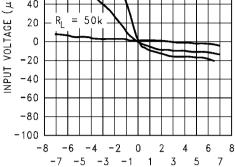
#### 60 VOLTAGE $(\mu V)$ 40 = 50k20 0 -20

Input Voltage vs. Output Voltage

R

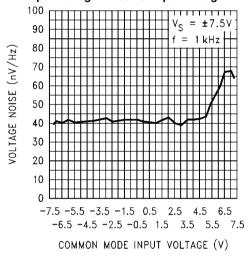
= 2k

= 600



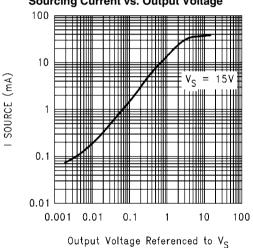
1199137

#### Input Voltage Noise vs. Input Voltage

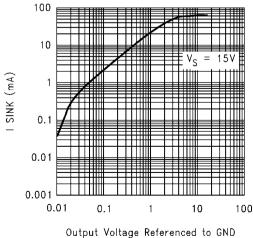


#### **Sourcing Current vs. Output Voltage**

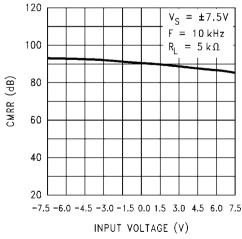
OUTPUT VOLTAGE (V)



Sinking Current vs. Output Voltage

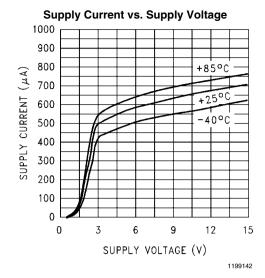


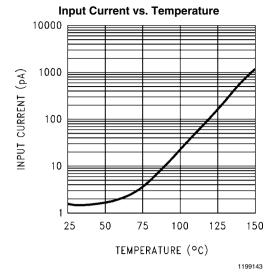
#### CMRR vs. Input Voltage

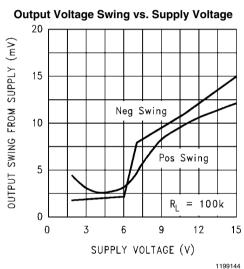


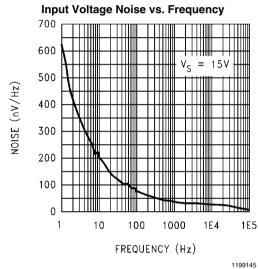
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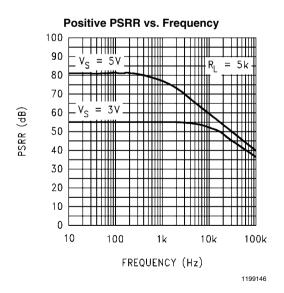
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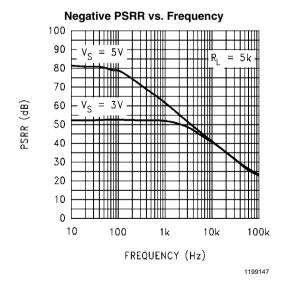


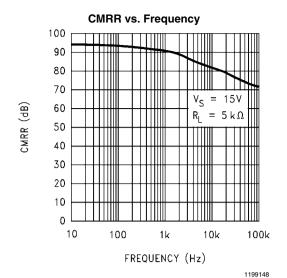


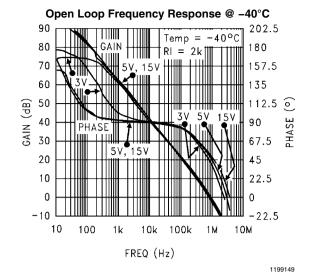












Open Loop Frequency Response @ 25°C

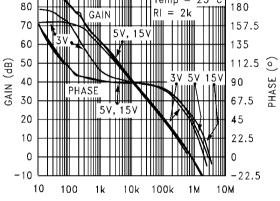
90

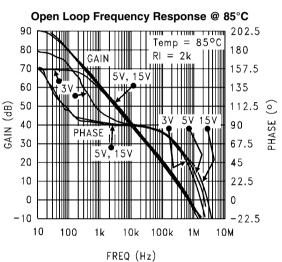
80

GAIN

RI = 2k

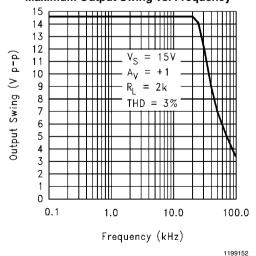
157.5

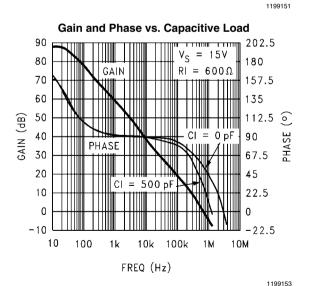




Maximum Output Swing vs. Frequency

FREQ (Hz)



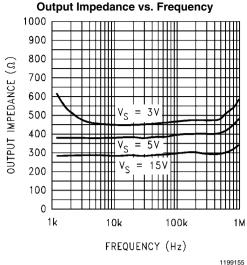


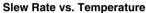
.....

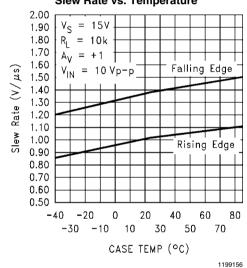
1199150

#### Gain and Phase vs. Capacitive Load 90 202.5 80 180 RI = 500kGAIN 70 157.5 60 135 50 GAIN (dB) 40 90 30 67.5 20 45 10 22.5 0 -10 22.5 10 100 100k 1 M 10M FREQ (Hz)

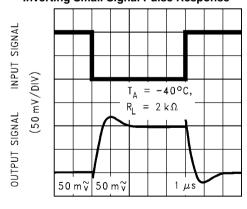
1199154







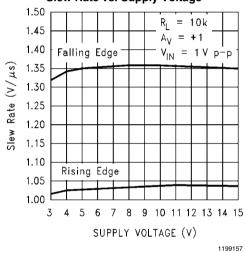
**Inverting Small Signal Pulse Response** 



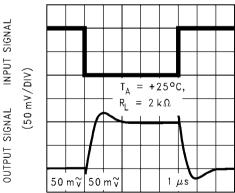
TIME  $(1\mu s/DIV)$ 

1199158

### Slew Rate vs. Supply Voltage



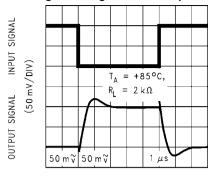
**Inverting Small Signal Pulse Response** 



TIME  $(1 \mu s/DIV)$ 

1199159

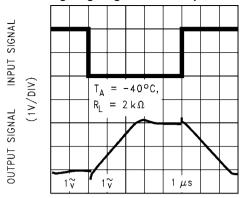
#### **Inverting Small Signal Pulse Response**



TIME (1  $\mu$ s/DIV)

1199160

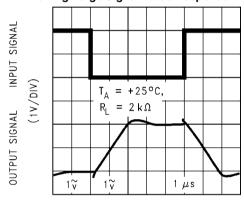
#### **Inverting Large Signal Pulse Response**



TIME  $(1 \mu s/DIV)$ 

1199161

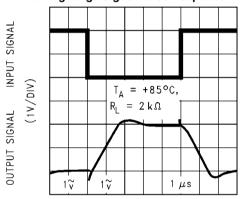
#### **Inverting Large Signal Pulse Response**



TIME  $(1 \mu s/DIV)$ 

1199162

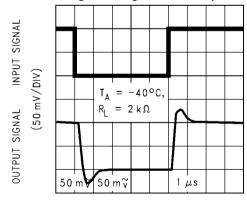
### **Inverting Large Signal Pulse Response**



TIME  $(1 \mu s/DIV)$ 

1199163

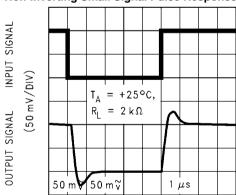
#### **Non-Inverting Small Signal Pulse Response**



TIME  $(1\mu s/DIV)$ 

1199164

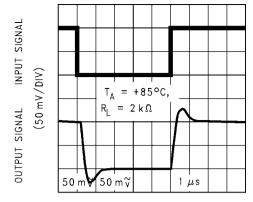
#### **Non-Inverting Small Signal Pulse Response**



TIME  $(1 \mu s/DIV)$ 

1199165

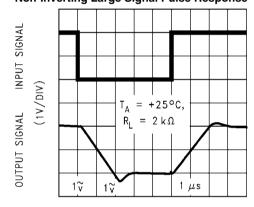
#### **Non-Inverting Small Signal Pulse Response**



TIME  $(1 \mu s/DIV)$ 

1199166

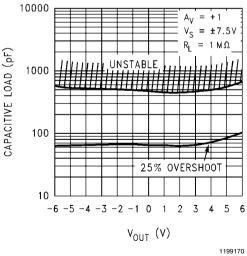
### Non-Inverting Large Signal Pulse Response



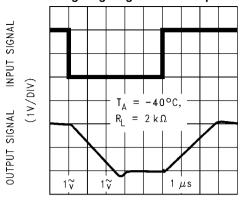
TIME  $(1 \mu s/DIV)$ 

1199168

# Stability vs. Capacitive Load



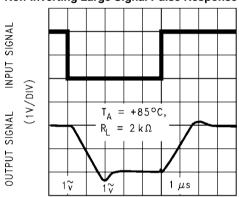
#### Non-Inverting Large Signal Pulse Response



TIME  $(1\mu s/DIV)$ 

1199167

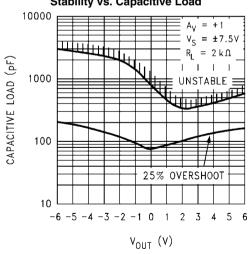
#### Non-Inverting Large Signal Pulse Response



TIME  $(1 \mu s/DIV)$ 

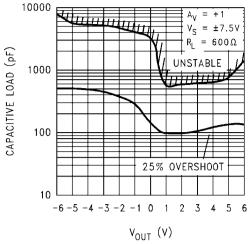
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#### Stability vs. Capacitive Load

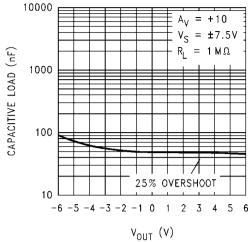


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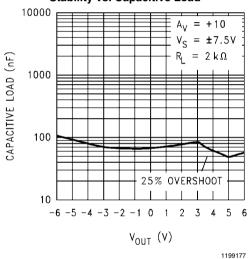
#### Stability vs. Capacitive Load



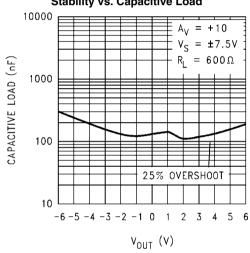
### Stability vs. Capacitive Load



#### Stability vs. Capacitive Load



### Stability vs. Capacitive Load



1199178

### **Application Information**

# 1.0 BENEFITS OF THE LMC7101 TINY AMP

#### Size

The small footprint of the SOT 23-5 packaged Tiny amp,  $(0.120 \times 0.118 \text{ inches}, 3.05 \times 3.00 \text{ mm})$  saves space on printed circuit boards, and enable the design of smaller electronic products. Because they are easier to carry, many customers prefer smaller and lighter products.

#### Height

The height (0.056 inches, 1.43 mm) of the Tiny amp makes it possible to use it in PCMCIA type III cards.

#### **Signal Integrity**

Signals can pick up noise between the signal source and the amplifier. By using a physically smaller amplifier package, the Tiny amp can be placed closer to the signal source, reducing noise pickup and increasing signal integrity. The Tiny amp can also be placed next to the signal destination, such as a buffer for the reference of an analog to digital converter.

#### **Simplified Board Layout**

The Tiny amp can simplify board layout in several ways. First, by placing an amp where amps are needed, instead of routing signals to a dual or quad device, long pc traces may be avoided.

By using multiple Tiny amps instead of duals or quads, complex signal routing and possibly crosstalk can be reduced.

#### **Low THD**

The high open loop gain of the LMC7101 amp allows it to achieve very low audio distortion—typically 0.01% at 10 kHz with a 10 k $\Omega$  load at 5V supplies. This makes the Tiny an excellent for audio, modems, and low frequency signal processing.

#### **Low Supply Current**

The typical 0.5 mA supply current of the LMC7101 extends battery life in portable applications, and may allow the reduction of the size of batteries in some applications.

#### **Wide Voltage Range**

The LMC7101 is characterized at 15V, 5V and 3V. Performance data is provided at these popular voltages. This wide voltage range makes the LMC7101 a good choice for devices where the voltage may vary over the life of the batteries.

#### 2.0 INPUT COMMON MODE

### **Voltage Range**

The LMC7101 does not exhibit phase inversion when an input voltage exceeds the negative supply voltage. *Figure 1* shows an input voltage exceeding both supplies with no resulting phase inversion of the output.

The absolute maximum input voltage is 300 mV beyond either rail at room temperature. Voltages greatly exceeding this maximum rating, as in *Figure 2*, can cause excessive current to flow in or out of the input pins, adversely affecting reliability.

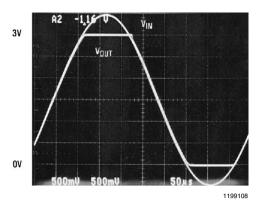


FIGURE 1. An Input Voltage Signal Exceeds the LMC7101 Power Supply Voltages with No Output Phase Inversion

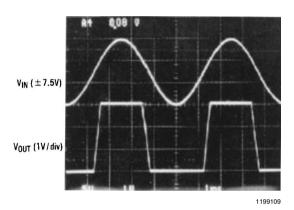


FIGURE 2. A ±7.5V Input Signal Greatly Exceeds the 3V Supply in *Figure 3* Causing No Phase Inversion Due to R<sub>1</sub>

Applications that exceed this rating must externally limit the maximum input current to ±5 mA with an input resistor as shown in *Figure 3*.

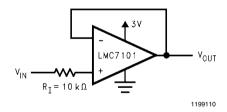


FIGURE 3. R<sub>I</sub> Input Current Protection for Voltages Exceeding the Supply Voltage

#### 3.0 RAIL-TO-RAIL OUTPUT

The approximate output resistance of the LMC7101 is  $180\Omega$  sourcing and  $130\Omega$  sinking at  $V_S=3V$  and  $110\Omega$  sourcing and  $80\Omega$  sinking at  $V_S=5V$ . Using the calculated output resistance, maximum output voltage swing can be estimated as a function of load.

#### 4.0 CAPACITIVE LOAD TOLERANCE

The LMC7101 can typically directly drive a 100 pF load with  $\rm V_S=15V$  at unity gain without oscillating. The unity gain follower is the most sensitive configuration. Direct capacitive loading reduces the phase margin of op amps. The combination of the op amp's output impedance and the capacitive load induces phase lag. This results in either an underdamped pulse response or oscillation.

Capacitive load compensation can be accomplished using resistive isolation as shown in *Figure 4*. This simple technique is useful for isolating the capacitive input of multiplexers and A/D converters.

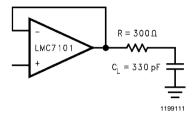


FIGURE 4. Resistive Isolation of a 330 pF Capacitive Load

# 5.0 COMPENSATING FOR INPUT CAPACITANCE WHEN USING LARGE VALUE FEEDBACK RESISTORS

When using very large value feedback resistors, (usually >  $500~\text{k}\Omega$ ) the large feed back resistance can react with the input capacitance due to transducers, photodiodes, and circuit board parasitics to reduce phase margins.

The effect of input capacitance can be compensated for by adding a feedback capacitor. The feedback capacitor (as in *Figure 5*),  $C_{\rm f}$  is first estimated by:

$$\frac{1}{2\pi R_1 \; C_{IN}} \geq \frac{1}{2\pi R_2 C_f}$$

or

$$R_1 C_{IN} \leq R_2 C_f$$

which typically provides significant overcompensation.

Printed circuit board stray capacitance may be larger or smaller than that of a breadboard, so the actual optimum value for  $C_{\rm F}$  may be different. The values of  $C_{\rm F}$  should be checked on the actual circuit. (Refer to the LMC660 quad CMOS amplifier data sheet for a more detailed discussion.)

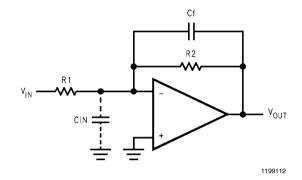


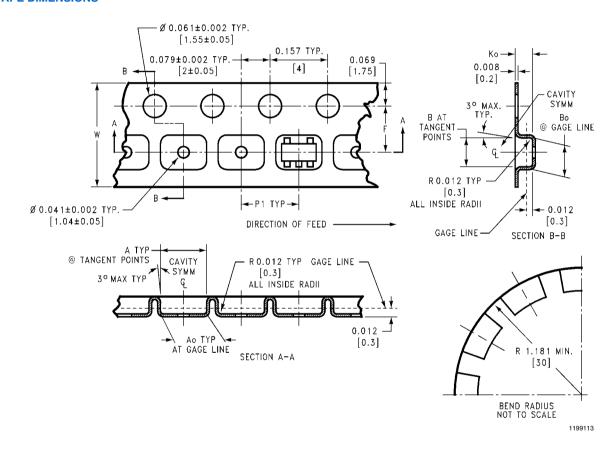
FIGURE 5. Cancelling the Effect of Input Capacitance

# **SOT23-5 Tape And Reel Specification**

#### **TAPE FORMAT**

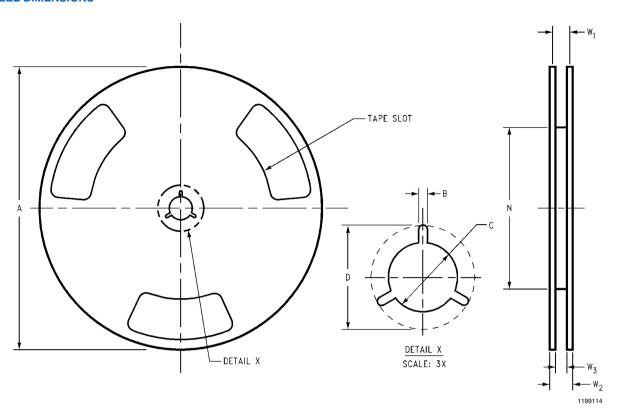
Tape Section	Tape Section # Cavities		Cover Tape Status
Leader	0 (min)	Empty	Sealed
(Start End)	75 (min)	Empty	Sealed
Commiss	3000	Filled	Sealed
Carrier	1000	Filled	Sealed
Trailer	125 (min)	Empty	Sealed
(Hub End)	0 (min)	Empty	Sealed

#### **TAPE DIMENSIONS**



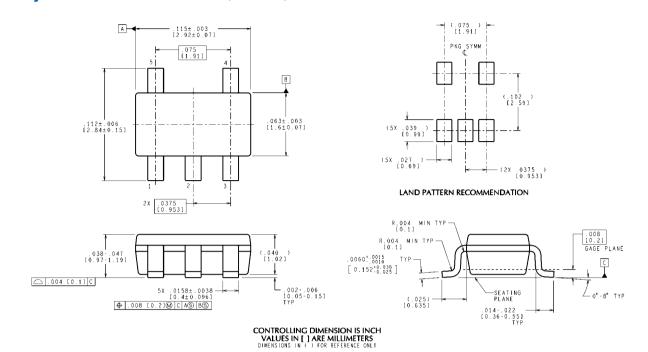
8 mm	0.130	0.124	0.130	0.126	0.138 ±0.002	0.055 ±0.004	0.157	0.315 ±0.012
	(3.3)	(3.15)	(3.3)	(3.2)	$(3.5 \pm 0.05)$	(1.4 ±0.11)	(4)	(8 ±0.3)
Tape Size	DIM A	DIM Ao	DIM B	DIM Bo	DIM F	DIM Ko	DIM P1	DIM W

### **REEL DIMENSIONS**



	9 mm	7.00	0.059	0.512	0.795	2.165	0.331 + 0.059/-0.000	0.567	W1+ 0.078/-0.039
8 mm 3:	330.00	1.50	13.00	20.20	55.00	8.40 + 1.50/-0.00	14.40	W1 + 2.00/-1.00	
	Tape Size	Α	В	С	D	N	W1	W2	W3

# Physical Dimensions inches (millimeters) unless otherwise noted



5-Pin SOT23 Package NS Package Number MF05A MF05A (Rev D)

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Data Converters	www.national.com/adc	Samples	www.national.com/samples		
Interface	www.national.com/interface	Eval Boards	www.national.com/evalboards		
LVDS	www.national.com/lvds	Packaging	www.national.com/packaging		
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Switching Regulators	www.national.com/switchers	Distributors	www.national.com/contacts		
LDOs	www.national.com/ldo	Quality and Reliability	www.national.com/quality		
LED Lighting	www.national.com/led	Feedback/Support	www.national.com/feedback		
Voltage Reference	www.national.com/vref	Design Made Easy	www.national.com/easy		
PowerWise® Solutions	www.national.com/powerwise	Solutions	www.national.com/solutions		
Serial Digital Interface (SDI)	www.national.com/sdi	Mil/Aero	www.national.com/milaero		
Temperature Sensors	www.national.com/tempsensors	SolarMagic™	www.national.com/solarmagic		
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